

For harmonic plane wave solutions $\vec{E} = E_\omega e^{i(\vec{k}\cdot\vec{r} - \omega t)}$
etc.

$$1) \Rightarrow i\vec{k} \cdot \vec{D}_\omega = i\vec{k} \cdot \epsilon_b E_\omega = 4\pi j_\omega = 4\pi \sigma \frac{\vec{k} \cdot \vec{E}_\omega}{\omega}$$

$$\Rightarrow i\vec{k} \cdot \vec{E}_\omega \left(\epsilon_b + \frac{4\pi i \sigma}{\omega} \right) = 0$$

$$2) \Rightarrow i\mu \vec{k} \cdot \vec{H}_\omega = 0$$

$$3) \Rightarrow i\vec{k} \times \vec{E}_\omega = \frac{i\omega \vec{B}_\omega}{c} = \frac{i\omega \mu \vec{H}_\omega}{c}$$

$$\begin{aligned} 4) \Rightarrow i\vec{k} \times \vec{H}_\omega &= \frac{4\pi}{c} \vec{j}_\omega - \frac{i\omega}{c} \vec{D}_\omega \\ &= \frac{4\pi}{c} \sigma \vec{E}_\omega - \frac{i\omega}{c} \epsilon_b \vec{E}_\omega \\ &= -\frac{i\omega}{c} \left(\epsilon_b + \frac{4\pi i \sigma}{\omega} \right) \vec{E}_\omega \end{aligned}$$

Notice: all the equations above look exactly like what we had for the dielectric, provided we define

$$\boxed{\epsilon(\omega) = \epsilon_b(\omega) + \frac{4\pi i \sigma(\omega)}{\omega}}$$

So all results for the dielectric case carry over to conductors, provided we make the above substitution. In particular

dispersion relation
for transverse modes

$$\boxed{k^2 = \frac{\omega^2}{c^2} \mu \epsilon(\omega)}$$

The main difference between dielectrics + conductors has to do with the contribution that the $4\pi i\sigma/\omega$ makes to the real and imaginary parts of $\epsilon(\omega)$.

For single Drude model $\sigma(\omega) = \frac{\sigma_0}{1 - i\omega\tau}$ $\sigma_0 = \frac{me^2}{m}$

① Low frequencies $\omega \ll 1/\tau$

$\epsilon_b(\omega) \approx \epsilon_b(0)$ real

$\sigma(\omega) \approx \sigma_0$ real

$\Rightarrow \boxed{\epsilon(\omega) \approx \epsilon_b(0) + \frac{4\pi i\sigma_0}{\omega}}$ ← gives large ϵ_2 as $\omega \rightarrow 0$

\Rightarrow strong dissipation

$\text{Re } \epsilon = \epsilon_1$

$\text{Im } \epsilon = \epsilon_2$

when $\frac{\epsilon_2}{\epsilon_1} = \frac{4\pi\sigma_0}{\omega\epsilon_b(0)} \gg 1$

we call this regime a "good" conductor.
conduction electrons dominate the response
- waves strongly attenuated

when $\frac{\epsilon_2}{\epsilon_1} = \frac{4\pi\sigma_0}{\omega\epsilon_b(0)} \ll 1$

we call this regime a "poor" conductor.
little absorption of energy by conduction electrons.
waves propagate

one always enters the "good" conductor region when ω gets sufficiently small.

wave vector;

$$k = \frac{\omega}{c} \sqrt{\mu \epsilon}$$

for a good conductor where $\epsilon_2 \gg \epsilon_1$,

$$\epsilon \sim i\epsilon_2 = \frac{4\pi i\sigma_0}{\omega}$$

$$k = k_1 + ik_2 = \frac{\omega}{c} \sqrt{\mu \frac{4\pi i\sigma_0}{\omega}} \quad \sqrt{i} = \frac{1+i}{\sqrt{2}}$$

$$k_1 = k_2 = \frac{\omega}{c} \sqrt{\frac{4\pi\mu\sigma_0}{2\omega}} = \frac{1}{c} \sqrt{2\pi\mu\sigma_0\omega}$$

for
 $\vec{k} = k\hat{z}$,

$$\vec{E} = \vec{E}_\omega e^{i(kz - \omega t)} = \vec{E}_\omega e^{-k_2 z} e^{i(k_1 z - \omega t)}$$

$$\delta \equiv 1/k_2 = \frac{c}{\sqrt{2\pi\mu\sigma_0\omega}}$$

"skin depth"
distance wave
propagates into
conductor

$$\delta \sim 1/\sqrt{\omega} \quad \text{increases as } \omega \text{ decreases}$$

ϕ phase shift between oscillations of \vec{E} and \vec{H}

$$\phi = \arctan(k_2/k_1) \approx \arctan(1) = 45^\circ$$

$$\text{Amplitude ratio } \frac{|\vec{H}_\omega|}{|\vec{E}_\omega|} = \frac{c|k|}{\omega\mu} = \frac{\sqrt{2}c}{\omega\mu} k_1$$

$$= \frac{\sqrt{2}c}{\omega\mu} \frac{1}{c} \sqrt{2\pi\mu\sigma_0\omega}$$

$$= \sqrt{\frac{4\pi\sigma_0}{\omega\mu}} \sim 1/\sqrt{\omega}$$

as $\omega \rightarrow 0$, most of the energy of the wave
is carried by the magnetic field part

② high frequencies $\omega \gg \frac{1}{\tau}$, $\omega \gg \omega_0$

$$\epsilon_b(\omega) \approx 1$$

$$\sigma(\omega) \approx \frac{\sigma_0}{-i\omega\tau} = \frac{ime^2\tau}{m\omega\tau} = \frac{ime^2}{m\omega}$$

pure imaginary
indep of τ

$$\epsilon(\omega) \approx 1 + \frac{4\pi i\sigma}{\omega} = 1 - \frac{4\pi me^2}{m\omega^2}$$

$$\boxed{\epsilon(\omega) = 1 - \frac{\omega_p^2}{\omega^2}}$$

$$\omega_p = \sqrt{\frac{4\pi me^2}{m}}$$

plasma freq of the
conduction electrons

$\epsilon(\omega)$ is real

1) If $\omega > \omega_p$ then $\epsilon > 0$

\Rightarrow transparent propagation

$$k = k_1 = \frac{\omega}{c} \sqrt{\mu\epsilon} \text{ is pure real}$$

$$k_2 \approx 0$$

2) If $\omega < \omega_p$ then $\epsilon < 0$

\Rightarrow total reflection

$$k_1 \approx 0$$

k is pure imaginary

$$k = k_2 = \frac{\omega}{c} \sqrt{\mu|\epsilon|}$$

ω_p gives cross over between total reflection
and transparent propagation

for typical metals

$$\tau \sim 10^{-14} \text{ sec}$$

$$\omega_p \sim 10^{16} \text{ sec}^{-1}$$

$$\lambda_p = \frac{2\pi c}{\omega_p} \sim 3 \times 10^3 \text{ \AA}$$

(visible is $\lambda \sim 5 \times 10^3 \text{ \AA}$)

Example: The ionosphere is a layer of charged gas surrounding the earth.

In many respects the charged particles of the ionosphere behave like conduction electrons in a metal. The plasma freq of the ionosphere is such that

for AM radio $\omega_{AM} < \omega_p \Rightarrow$ AM radio signals reflected back to earth

for FM radio $\omega_{FM} > \omega_p \Rightarrow$ FM radio signals propagate through ionosphere into space

Explains why you can pick up AM stations from far away - they get reflected back
But you can only pick up local FM stations.

Longitudinal modes in conductors

ie \vec{H}_ω or \vec{E}_ω not $\perp \vec{k}$
magnetic field

$$\vec{\nabla} \cdot \vec{B} = 0 \Rightarrow i\mu \vec{k} \cdot \vec{H}_\omega = 0 \Rightarrow \vec{H}_\omega \perp \vec{k} \text{ transverse}$$

or $\vec{k} = 0$ spatially uniform \vec{H}

if $\vec{k} = 0$ then Faraday

$$i\vec{k} \times \vec{E}_\omega = i\omega\mu \vec{H}_\omega = 0 \Rightarrow \omega = 0$$

" as $\vec{k} = 0$

So only possible longitudinal \vec{H} is spatially uniform, constant in time.

electric field

$$\vec{\nabla} \cdot \vec{D} = 4\pi\rho_f \Rightarrow i\varepsilon(\omega) \vec{k} \cdot \vec{E}_\omega = 0 \Rightarrow \vec{E}_\omega \perp \vec{k} \text{ transverse}$$

or $\varepsilon(\omega) = 0$

If $\vec{E}_\omega \parallel \vec{k}$ but $\varepsilon(\omega) = 0$, then can satisfy all other Maxwell equations.

$$i\vec{k} \times \vec{E}_\omega = \frac{i\omega\mu}{c} \vec{H}_\omega \Rightarrow \vec{H}_\omega = 0$$

$$\Rightarrow i\rho_0 \vec{k} \cdot \vec{H}_\omega = 0 \quad \text{and} \quad i\vec{k} \times \vec{H}_\omega = -\frac{i\omega\varepsilon(\omega)}{c} \vec{E}_\omega$$

" as $\vec{H}_\omega = 0$ " as $\varepsilon(\omega) = 0$

So we can have longitudinal electric field oscillation when $\varepsilon(\omega) = 0$

low freq $\omega \ll \omega_0$, $\omega \tau \ll 1$

$$\epsilon \approx \epsilon_b(\omega) + \frac{4\pi i \sigma_0}{\omega}$$

$$\epsilon(\omega) = 0 \quad \text{when} \quad \omega = -\frac{4\pi i \sigma_0}{\epsilon_b(\omega)}$$

$$\vec{E}(\vec{r}, t) = \vec{E}_\omega e^{i(\vec{k} \cdot \vec{r} - \omega t)} = \vec{E}_\omega e^{-\frac{4\pi \sigma_0}{\epsilon_b(\omega)} t} e^{i\vec{k} \cdot \vec{r}}$$

If set up a longitudinal \vec{E} field, it decays to zero exponentially with ~~time~~ decay time $\epsilon_b(\omega)/4\pi\sigma_0$. This is consistent with assumption that $\vec{E} = 0$ inside a conductor for electrostatics.

in statics $\vec{E} = -\vec{\nabla}\phi \Rightarrow \vec{E} \sim -i\vec{k}\phi_k e^{i\vec{k} \cdot \vec{r}}$ is longitudinal.

high freq $\omega \gg 1/\tau$, $\omega \gg \omega_0$

$$\epsilon(\omega) \approx 1 + \frac{4\pi i \sigma_0}{\omega} = 1 - \frac{\omega_p^2}{\omega^2} \quad \omega_p^2 = \frac{4\pi m e^2}{m}$$

$$\epsilon = 0 \quad \text{when} \quad \omega = \omega_p$$

So we have oscillatory longitudinal \vec{E} only when $\omega = \omega_p$, independent of \vec{k} .

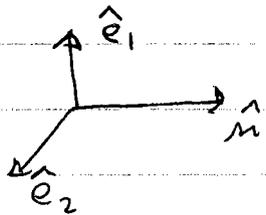
$$\vec{E} = \vec{E}_\omega e^{i\vec{k} \cdot \vec{r}} e^{-i\omega_p t}$$

This is called a plasma oscillation. When one quantizes this oscillatory mode, it is called a plasmon.

$$\vec{\nabla} \cdot \vec{E} = 4\pi \rho \Rightarrow \rho = \frac{i\vec{k} \cdot \vec{E}_\omega}{4\pi} e^{i\vec{k} \cdot \vec{r}} e^{-i\omega_p t} \left\{ \begin{array}{l} \text{plasma osc.} \\ \text{is a charge} \\ \text{density oscillation} \end{array} \right.$$

Polarization

Consider transverse wave propagating in direction \hat{m}
i.e. $\vec{k} = k \hat{m}$.



$$\begin{aligned}\hat{e}_1 \times \hat{e}_2 &= \hat{m} \\ \hat{m} \times \hat{e}_1 &= \hat{e}_2 \\ \hat{e}_2 \times \hat{m} &= \hat{e}_1\end{aligned}$$

A general solution for a transverse wave has the form

$$\vec{E}(\vec{r}, t) = \text{Re} \left\{ (E_1 \hat{e}_1 + E_2 \hat{e}_2) e^{i(\vec{k} \cdot \vec{r} - \omega t)} \right\}$$

$$\begin{aligned}\vec{H}(\vec{r}, t) &= \frac{c}{\omega \mu} \text{Re} \left\{ k \hat{m} \times (E_1 \hat{e}_1 + E_2 \hat{e}_2) e^{i(\vec{k} \cdot \vec{r} - \omega t)} \right\} \\ &= \frac{c}{\mu \omega} \text{Re} \left\{ k (E_1 \hat{e}_2 - E_2 \hat{e}_1) e^{i(\vec{k} \cdot \vec{r} - \omega t)} \right\}\end{aligned}$$

So far we considered implicitly only the case

$$E_1, E_2 \text{ real constants} \Rightarrow |\vec{E}_\omega| = \sqrt{E_1^2 + E_2^2}$$

But most general case is

$$E_1 = E \cos \theta e^{i\chi_1}$$

$$E_2 = E \sin \theta e^{i\chi_2}$$

$$E^2 = |E_1|^2 + |E_2|^2$$

} need not be real
can be complex with
relative phase difference

$$\text{define } \Phi = k_1 \hat{m} \cdot \vec{r} - \omega t$$

$$\tan \Phi = k_2 / k_1$$

(For $\chi_1 = \chi_2 = 0$, θ
is angle \vec{E}_ω makes
with respect to \hat{e}_1)

$$\vec{E} = E e^{-k_2 \hat{m} \cdot \vec{r}} \left[\hat{e}_1 \cos \theta \cos(\Phi + \chi_1) + \hat{e}_2 \sin \theta \cos(\Phi + \chi_2) \right]$$

$$\vec{H} = \frac{ck}{\omega \mu} E e^{-k_2 \hat{m} \cdot \vec{r}} \left[\hat{e}_2 \cos \theta \cos(\Phi + \chi_1 + \varphi) - \hat{e}_1 \sin \theta \cos(\Phi + \chi_2 + \varphi) \right]$$

special cases

1) $\chi_1 = \chi_2$ $\vec{E} = (\hat{e}_1 \cos \theta + \hat{e}_2 \sin \theta) E e^{-k_2 \hat{m} \cdot \vec{r}} \cos(\Phi + \chi)$
 $\vec{H} = (-\hat{e}_1 \sin \theta + \hat{e}_2 \cos \theta) \frac{ck}{\omega \mu} E e^{-k_2 \hat{m} \cdot \vec{r}} \cos(\Phi + \chi + \varphi)$
Linearly polarized - \vec{E} and \vec{H} point in fixed directions and are orthogonal. phase shift is φ

2) $\cos \theta = \frac{1}{\sqrt{2}}, \sin \theta = \pm \frac{1}{\sqrt{2}}$, $\chi_1 = 0$ (can always choose $\chi_1 = 0$ by shifting the time scale)
 $\chi_2 = \chi$

$$\vec{E}^\pm = \frac{E}{\sqrt{2}} (\hat{e}_1 \cos \Phi \pm \hat{e}_2 \cos(\Phi + \chi))$$

Find locus of points that \vec{E} sweeps out as Φ varies.

$$\begin{aligned} \vec{E}^\pm \cdot \hat{e}_1 &= E_1^\pm = \frac{E}{\sqrt{2}} \cos \Phi \\ \vec{E}^\pm \cdot \hat{e}_2 &= E_2^\pm = \pm \frac{E}{\sqrt{2}} \cos(\Phi + \chi) \end{aligned}$$

Define $\Theta = \Phi + \chi/2$ so $E_1^\pm = \frac{E}{\sqrt{2}} \cos(\Theta - \chi/2)$
 $E_2^\pm = \pm \frac{E}{\sqrt{2}} \cos(\Theta + \chi/2)$

$$\begin{aligned} E_1^\pm &= \frac{E}{\sqrt{2}} \cos \Theta \cos \chi/2 + \frac{E}{\sqrt{2}} \sin \Theta \sin \chi/2 \\ E_2^\pm &= \pm \frac{E}{\sqrt{2}} \cos \Theta \cos \chi/2 \mp \frac{E}{\sqrt{2}} \sin \Theta \sin \chi/2 \end{aligned}$$

$$\Rightarrow \begin{aligned} E_1^+ + E_2^+ &= \frac{2E \cos \theta \cos \chi/2}{\sqrt{2}} \\ E_1^+ - E_2^+ &= \frac{2E \sin \theta \sin \chi/2}{\sqrt{2}} \end{aligned}$$

$$\Rightarrow \frac{(E_1^+ + E_2^+)^2}{2E^2 \cos^2 \chi/2} + \frac{(E_1^+ - E_2^+)^2}{2E^2 \sin^2 \chi/2} = \cos^2 \theta + \sin^2 \theta = 1$$

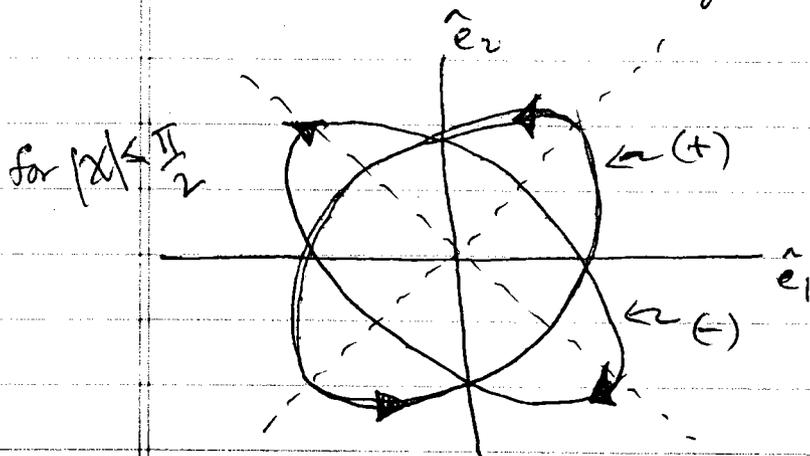
Similarly

$$\frac{(E_1^- - E_2^-)^2}{2E^2 \cos^2 \chi/2} + \frac{(E_1^- + E_2^-)^2}{2E^2 \sin^2 \chi/2} = 1$$

These are the equations for ellipses! $\frac{x^2}{a^2} + \frac{y^2}{b^2} = 1$
with semi axes

$E \cos \chi/2$ and $E \sin \chi/2$

direction of the ellipse axes are $\left(\frac{\hat{e}_1 + \hat{e}_2}{\sqrt{2}}\right)$ and $\left(\frac{\hat{e}_1 - \hat{e}_2}{\sqrt{2}}\right)$



$\vec{E} \cdot \hat{e}_1 = \frac{E_1 + E_2}{\sqrt{2}}$, $\vec{E} \cdot \hat{e}_2 = \frac{E_1 - E_2}{\sqrt{2}}$
so $|\vec{E}|$ sweeps out ellipse as Φ varies.

\Rightarrow sit at position \vec{r}
as t varies, $|\vec{E}|$
sweeps out ellipse
axis at 45° to \hat{e}_1, \hat{e}_2

elliptically polarized wave

For $0 < \chi < \frac{\pi}{2}$

For (+) \vec{E} moves around ellipse counterclockwise (right handed)
For (-) \vec{E} moves around ellipse clockwise (left handed)
as t increases (i.e. as Φ decreases)

Special case of 1v) $\chi = \pi/2$

$$\cos^2 \chi/2 = \sin^2 \chi/2 = \frac{1}{2} \quad \text{ellipse axes are equal!}$$

$\Rightarrow \vec{E}$ sweeps out a circular path

circularly polarized waves

(+) goes counterclockwise

(-) goes clockwise

One defines circular polarization basis vectors as:

$$\hat{e}_{\pm} \equiv \frac{1}{\sqrt{2}} (\hat{e}_1 \pm i\hat{e}_2)$$

$$\Rightarrow \hat{e}_{\pm}^* \cdot \hat{e}_{\pm} = \hat{e}_{\pm} \cdot \hat{e}_{\mp} = 1$$

$$\hat{e}_{\pm} \cdot \hat{e}_{\mp}^* = \hat{e}_{\pm} \cdot \hat{e}_{\pm} = 0$$

$$\hat{e}_{\pm} \cdot \hat{m} = 0$$

$$\hat{m} \times \hat{e}_{\pm} = i\hat{e}_{\pm}$$

With this notation, a circularly polarized wave is

$$\vec{E} = \text{Re} \left\{ E \hat{e}_{\pm} e^{i(\vec{k} \cdot \vec{r} - \omega t)} \right\} \quad \begin{array}{l} (+) \text{ is counterclockwise} \\ (-) \text{ is clockwise} \end{array}$$

Note:

with \hat{m} pointing out

$$\begin{aligned} \frac{1}{\sqrt{2}} (E \hat{e}_+ + E \hat{e}_-) &= \frac{E}{\sqrt{2}} \left(\frac{1}{\sqrt{2}} \hat{e}_1 + \frac{i}{\sqrt{2}} \hat{e}_2 + \frac{1}{\sqrt{2}} \hat{e}_1 - \frac{i}{\sqrt{2}} \hat{e}_2 \right) \\ &= E \hat{e}_1 \end{aligned}$$

$$\text{and } \frac{1}{\sqrt{2}i} (E \hat{e}_+ - E \hat{e}_-) = E \hat{e}_2$$

Thus a linearly polarized wave can be written as a superposition of counter rotating circularly polarized waves!

general case E_1, E_2 complex

$$\text{write } E_1 \hat{e}_1 + E_2 \hat{e}_2 \equiv \vec{U} e^{i\psi}$$

where ψ is chosen so that $\vec{U} \cdot \vec{U}$ is real
(can always do this since $\vec{U} \cdot \vec{U} = (E_1^2 + E_2^2) e^{-2i\psi}$
so 2ψ is just the phase of $E_1^2 + E_2^2$)

\vec{U} is complex vector $\Rightarrow \vec{U} = \vec{U}_a - i\vec{U}_b$, \vec{U}_a, \vec{U}_b real
since $\vec{U} \cdot \vec{U}$ is real $\Rightarrow \vec{U}_a \cdot \vec{U}_b = 0$
so $\vec{U}_a \perp \vec{U}_b$ orthogonal

$$\begin{aligned} \vec{E}(\vec{r}, t) &= \text{Re} \left\{ \vec{U} e^{i\psi} e^{i(\vec{k} \cdot \vec{r} - \omega t)} \right\} & \Phi &= \vec{k} \cdot \vec{r} - \omega t \\ &= e^{-k_2 \hat{m} \cdot \vec{r}} \left\{ \vec{U}_a \cos(\Phi + \psi) + \vec{U}_b \sin(\Phi + \psi) \right\} \end{aligned}$$

Define E_a as component of \vec{E} in direction of \vec{U}_a
 E_b as component of \vec{E} in direction of \vec{U}_b

$$E_a = U_a \cos(\Phi + \psi)$$

$$E_b = U_b \sin(\Phi + \psi)$$

(ignore attenuation factor by either absorbing it into U_a, U_b , or consider $\vec{r} = 0$)

$$\Rightarrow \left(\frac{E_a}{U_a} \right)^2 + \left(\frac{E_b}{U_b} \right)^2 = 1$$

$$\begin{aligned} U_a &\equiv |\vec{U}_a| \\ U_b &\equiv |\vec{U}_b| \end{aligned}$$

elliptical polarization

semi axes of lengths U_a and U_b
oriented in directions \vec{U}_a and \vec{U}_b

if $U_a = \pm U_b$ then circularly polarized

Define circular polarization basis vectors

$$\hat{e}_+ = \frac{1}{\sqrt{2}} (\hat{e}_a + i\hat{e}_b)$$

$$\hat{e}_a = \frac{u_a}{|u_a|}, \quad \hat{e}_b = \frac{u_b}{|u_b|}$$

any general \vec{u} can always be written as:

$$\vec{u} = \frac{1}{\sqrt{2}} (u_a + u_b) \hat{e}_- + \frac{1}{\sqrt{2}} (u_a - u_b) \hat{e}_+$$

Thus a general elliptically polarized wave can be written as a superposition of circularly polarized waves!