

For harmonic plane wave solutions $\vec{E} = E_{\omega} e^{i(\vec{k} \cdot \vec{r} - \omega t)}$
etc.

$$1) \Rightarrow i\vec{k} \cdot \vec{D}_{\omega} = i\vec{k} \cdot \epsilon_b \vec{E}_{\omega} = 4\pi f_{\omega} = 4\pi \sigma \frac{\vec{k} \cdot \vec{E}_{\omega}}{\omega}$$

$$\Rightarrow i\vec{k} \cdot \vec{E}_{\omega} \left(\epsilon_b + \frac{4\pi i \sigma}{\omega} \right) = 0$$

$$2) \Rightarrow i\mu \vec{k} \cdot \vec{H}_{\omega} = 0$$

$$3) \Rightarrow i\vec{k} \times \vec{E}_{\omega} = \frac{i\omega}{c} \vec{B}_{\omega} = \frac{i\omega\mu}{c} \vec{H}_{\omega}$$

$$\begin{aligned} 4) \Rightarrow i\vec{k} \times \vec{H}_{\omega} &= \frac{4\pi}{c} \vec{f}_{\omega} - \frac{i\omega}{c} \vec{D}_{\omega} \\ &= \frac{4\pi\sigma}{c} \vec{E}_{\omega} - \frac{i\omega}{c} \epsilon_b \vec{E}_{\omega} \\ &= -\frac{i\omega}{c} \left(\epsilon_b + \frac{4\pi i \sigma}{\omega} \right) \vec{E}_{\omega} \end{aligned}$$

Notice: all the equations above look exactly like what we had for the dielectric, provided we define

$$\boxed{\epsilon(\omega) = \epsilon_b(\omega) + \frac{4\pi i \sigma(\omega)}{\omega}}$$

So all results for the dielectric case carry over to conductors, provided we make the above substitution. In particular

dispersion relation
for transverse modes

$$\boxed{k^2 = \frac{\omega^2}{c^2} \mu \epsilon(\omega)}$$

The main difference between dielectrics + conductors has to do with the contribution that the $4\pi i\sigma/\omega$ makes to the real and imaginary parts of $\epsilon(\omega)$.

For single Drude model $\sigma(\omega) = \frac{\sigma_0}{1 - i\omega\tau}$ $\sigma_0 = \frac{ne^2\tau}{m}$

① Low frequencies $\omega \ll 1/\tau$

$$\epsilon_b(\omega) \approx \epsilon_b(0) \text{ real}$$

$$\sigma(\omega) \approx \sigma_0 \text{ real}$$

$$\Rightarrow \boxed{\epsilon(\omega) \approx \epsilon_b(0) + \frac{4\pi i\sigma_0}{\omega}} \leftarrow \begin{array}{l} \text{gives large } \epsilon_2 \text{ as} \\ \omega \rightarrow 0 \end{array}$$

$\Rightarrow$ strong dissipation

$$\text{Re } \epsilon = \epsilon_1$$

$$\text{Im } \epsilon = \epsilon_2$$

when $\frac{\epsilon_2}{\epsilon_1} = \frac{4\pi\sigma_0}{\omega\epsilon_b(0)} \gg 1$ we call this regime a "good" conductor.

Conduction electrons dominate the response
- waves strongly attenuated

when $\frac{\epsilon_2}{\epsilon_1} = \frac{4\pi\sigma_0}{\omega\epsilon_b(0)} \ll 1$ we call this regime a "poor" conductor.

little absorption of energy by conduction electrons.
waves propagate

one always enters the "good" conductor region when ω gets sufficiently small.

wave vector:

$$k = \frac{\omega}{c} \sqrt{\mu \epsilon}$$

for a good conductor where $\epsilon_2 \gg \epsilon_1$,

$$\epsilon \sim i\epsilon_2 = \frac{4\pi i\sigma_0}{\omega}$$

$$k = k_1 + ik_2 = \frac{\omega}{c} \sqrt{\mu \frac{4\pi i\sigma_0}{\omega}} \quad \sqrt{i} = \frac{1+i}{\sqrt{2}}$$

$$k_1 = k_2 = \frac{\omega}{c} \sqrt{\frac{4\pi\mu\sigma_0}{2\omega}} = \frac{1}{c} \sqrt{2\pi\mu\sigma_0\omega}$$

for $\vec{k} = k\hat{z}$,

$$\vec{E} = \vec{E}_0 e^{-i(kz - \omega t)} = \vec{E}_0 e^{-k_2 z} e^{i(k_1 z - \omega t)}$$

$$\delta \equiv 1/k_2 = \frac{c}{\sqrt{2\pi\mu\sigma_0\omega}} \quad \text{"skin depth"}$$

$\delta \sim 1/\sqrt{\omega}$ increases as ω decreases
distance wave propagates into conductor

ϕ phase shift between oscillations of $\vec{E}$ and $\vec{H}$

$$\phi = \arctan(k_2/k_1) \approx \arctan(1) = 45^\circ$$

$$\text{Amplitude ratio } \frac{|\vec{H}_0|}{|\vec{E}_0|} = \frac{c|k|}{\omega\mu} = \frac{\sqrt{2}c}{\omega\mu} k_1$$

$$= \frac{\sqrt{2}c}{\omega\mu} \frac{1}{c} \sqrt{2\pi\mu\sigma_0\omega}$$

$$= \sqrt{\frac{4\pi\sigma_0}{\omega\mu}} \sim 1/\sqrt{\omega}$$

as $\omega \rightarrow 0$, most of the energy of the wave is carried by the magnetic field part

② high frequencies $\omega \gg 1/\tau$, $\omega \gg \omega_0$

$$\epsilon_b(\omega) \approx 1$$

$$\sigma(\omega) \approx \frac{\sigma_0}{-i\omega\tau} = \frac{ime^2\tau}{m\omega\tau} = \frac{ime^2}{m\omega}$$

pure imaginary
indep of τ

$$\epsilon(\omega) \approx 1 + \frac{4\pi i\sigma}{\omega} \approx 1 - \frac{4\pi me^2}{m\omega^2}$$

$$\boxed{\epsilon(\omega) \approx 1 - \frac{\omega_p^2}{\omega^2}}$$

$$\omega_p \equiv \sqrt{\frac{4\pi me^2}{m}}$$

plasma freq of the
conduction electrons

$\epsilon(\omega)$ is real

1) If $\omega > \omega_p$ then $\epsilon > 0$

$\Rightarrow$ transparent propagation

$$k = k_1 = \frac{\omega}{c} \sqrt{\mu\epsilon} \text{ is pure real}$$

$$k_2 \approx 0$$

2) If $\omega < \omega_p$ then $\epsilon < 0$

$\Rightarrow$ total reflection

$$k_1 \approx 0$$

k is pure imaginary

$$k = k_2 = \frac{\omega}{c} \sqrt{\mu|\epsilon|}$$

ω_p gives cross over between total reflection
and transparent propagation

for typical metals

$$\tau \sim 10^{-14} \text{ sec}$$

$$\omega_p \sim 10^{16} \text{ sec}^{-1}$$

$$\lambda_p = \frac{2\pi c}{\omega_p} \sim 3 \times 10^3 \text{ \AA} \quad (\text{visible is } \lambda \sim 5 \times 10^3 \text{ \AA})$$

Example: The ionosphere is a layer of charged gas surrounding the earth.

In many respects the charged particles of the ionosphere behave like conduction electrons in a metal. The plasma freq of the ionosphere is such that

for AM radio $\omega_{AM} < \omega_p \Rightarrow$ AM radio signals reflected back to earth

for FM radio $\omega_{FM} > \omega_p \Rightarrow$ FM radio signals propagate through ionosphere into space

Explains why you can pick up AM stations from far away - they get reflected back. But you can only pick up local FM stations.

Longitudinal modes in conductors

is $\vec{H}_\omega$ or $\vec{E}_\omega$ not $\perp \vec{k}$

magnetic field

$$\vec{\nabla} \cdot \vec{B} = 0 \Rightarrow \vec{\nabla} \cdot \vec{H} = 0 \Rightarrow \vec{H}_\omega \perp \vec{k} \text{ transverse}$$

or $\vec{k} = 0$ spatially uniform $\vec{H}$

if $\vec{k} = 0$ then Faraday

$$\vec{k} \times \vec{E}_\omega = i\omega \mu \vec{H}_\omega = 0 \Rightarrow \omega = 0$$

as $\vec{k} = 0$

So only possible longitudinal $\vec{H}$ is

spatially uniform, constant in time.

electric field

$$\vec{\nabla} \cdot \vec{D} = 4\pi q_+ \Rightarrow \vec{\nabla} \cdot \vec{E}_\omega = 0 \Rightarrow \vec{E}_\omega \perp \vec{k} \text{ transverse}$$

$$\text{or } \vec{E}_\omega = 0$$

If $\vec{E}_\omega \parallel \vec{k}$ but $\vec{E}_\omega = 0$, then can satisfy all other Maxwell equations.

$$\vec{k} \times \vec{E}_\omega = i\omega \mu \vec{H}_\omega \Rightarrow \vec{H}_\omega = 0$$

$$\Rightarrow \vec{k} \cdot \vec{H}_\omega = 0 \text{ and } \vec{k} \times \vec{H}_\omega = -i\omega \vec{E}_\omega \Rightarrow \vec{E}_\omega = 0$$

as $\vec{H}_\omega = 0$ as $\vec{E}_\omega = 0$

So we can have longitudinal electric field oscillation

$$\text{or } \vec{E}_\omega = 0$$

low freq $\omega \ll \omega_0$ $\omega \tau \ll 1$

$$\epsilon \approx \epsilon_b(0) + \frac{4\pi i \sigma_0}{\omega}$$

$$\epsilon(\omega) = 0 \quad \text{when} \quad \omega = -\frac{4\pi i \sigma_0}{\epsilon_b(0)}$$

$$\vec{E}(\vec{r}, t) = \vec{E}_\omega e^{i(\vec{k} \cdot \vec{r} - \omega t)} = \vec{E}_\omega e^{-\frac{4\pi \sigma_0}{\epsilon_b(0)} t} e^{i\vec{k} \cdot \vec{r}}$$

If set up a longitudinal $\vec{E}$ field, it decays to zero exponentially with ~~time~~ decay time $\epsilon_b(0)/4\pi\sigma_0$. This is consistent with assumption that $\vec{E}=0$ inside a conductor for electrostatics.

in statics $\vec{E} = -\vec{\nabla}\phi \Rightarrow \vec{E} \sim -i\vec{k}\phi_k e^{i\vec{k} \cdot \vec{r}}$ is longitudinal

high freq $\omega \gg 1/\tau$, $\omega \gg \omega_0$

$$\epsilon(\omega) \approx 1 + \frac{4\pi i \sigma_0}{\omega} = 1 - \frac{\omega_p^2}{\omega^2} \quad \omega_p^2 = \frac{4\pi m e^2}{m}$$

$$\epsilon = 0 \quad \text{when} \quad \omega = \omega_p$$

So we have oscillatory longitudinal $\vec{E}$ only when $\omega = \omega_p$, independent of $\vec{k}$.

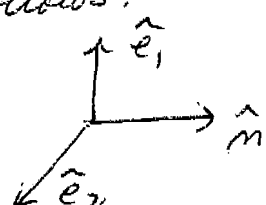
$$\vec{E} = \vec{E}_\omega e^{i\vec{k} \cdot \vec{r}} e^{-i\omega_p t}$$

This is called a plasma oscillation. When one quantizes this oscillatory mode, it is called a plasmon.

$$\vec{\nabla} \cdot \vec{E} = 4\pi \rho \Rightarrow \rho = \frac{i\vec{k} \cdot \vec{E}_\omega}{4\pi} e^{i\vec{k} \cdot \vec{r}} e^{-i\omega_p t} \left\{ \begin{array}{l} \text{plasma osc.} \\ \text{is a charge} \\ \text{density oscillation} \end{array} \right.$$

Polarization

Consider a transverse plane wave traveling in direction $\hat{m}$, i.e. $\vec{k} = k \hat{m}$. Define a right-handed coordinate system as follows:



$$\begin{aligned}\hat{e}_1 \times \hat{e}_2 &= \hat{m} \\ \hat{m} \times \hat{e}_1 &= \hat{e}_2 \\ \hat{e}_2 \times \hat{m} &= \hat{e}_1\end{aligned}$$

A general solution to Maxwell's equations for a transverse plane wave is then

$$\begin{aligned}\vec{E}(\vec{r}, t) &= \text{Re} \left\{ (E_1 \hat{e}_1 + E_2 \hat{e}_2) e^{i(\vec{k} \cdot \vec{r} - \omega t)} \right\} \\ \vec{H}(\vec{r}, t) &= \frac{c}{\omega \mu} \text{Re} \left\{ k \hat{m} \times (E_1 \hat{e}_1 + E_2 \hat{e}_2) e^{i(\vec{k} \cdot \vec{r} - \omega t)} \right\} \\ &= \frac{c}{\omega \mu} \text{Re} \left\{ k (E_1 \hat{e}_2 - E_2 \hat{e}_1) e^{i(\vec{k} \cdot \vec{r} - \omega t)} \right\}\end{aligned}$$

In general, k is complex
 $k = k_1 + i k_2 = |k| e^{i\delta}, \quad \begin{cases} |k| = \sqrt{k_1^2 + k_2^2} \\ \delta = \arctan(k_2/k_1) \end{cases}$

So far we implicitly assumed that E_1 and E_2 are real constants. In this case

$$\begin{aligned}\vec{E}(\vec{r}, t) &= \vec{E}_0 e^{-k_2 \hat{m} \cdot \vec{r}} \cos(k_1 \hat{m} \cdot \vec{r} - \omega t) \\ \vec{H}(\vec{r}, t) &= \vec{H}_0 e^{-k_2 \hat{m} \cdot \vec{r}} \cos(k_1 \hat{m} \cdot \vec{r} - \omega t + \delta)\end{aligned}$$

where

$$\vec{E}_0 \equiv E_1 \hat{e}_1 + E_2 \hat{e}_2 \quad \text{and} \quad \vec{H}_0 \equiv \frac{c|k|}{\omega \mu} (E_1 \hat{e}_2 - E_2 \hat{e}_1)$$

are fixed vectors for all time and space.

In this case the directions of $\vec{E}$ and $\vec{H}$ remain fixed while the amplitudes oscillate in time and space. Such a plane wave is called a linearly polarized wave.

However there is nothing to prevent one from choosing a solution with E_1 and E_2 complex numbers,

$$E_1 = |E_1| e^{i\chi_1}, \quad E_2 = |E_2| e^{i\chi_2}$$

In this case one has

$$\begin{aligned} \vec{E}(\vec{r}, t) &= \text{Re} \left\{ |E_1| \hat{e}_1 e^{i(\vec{k} \cdot \vec{r} - \omega t + \chi_1)} + |E_2| \hat{e}_2 e^{i(\vec{k} \cdot \vec{r} - \omega t + \chi_2)} \right\} \\ &= e^{-k_2 \hat{m} \cdot \vec{r}} \left[|E_1| \hat{e}_1 \cos(k_1 \hat{m} \cdot \vec{r} - \omega t + \chi_1) \right. \\ &\quad \left. + |E_2| \hat{e}_2 \cos(k_1 \hat{m} \cdot \vec{r} - \omega t + \chi_2) \right] \end{aligned}$$

and

$$\begin{aligned} \vec{H}(\vec{r}, t) &= \frac{c|k|}{\omega\mu} \text{Re} \left\{ |E_1| \hat{e}_2 e^{i(\vec{k} \cdot \vec{r} - \omega t + \delta + \chi_1)} \right. \\ &\quad \left. - |E_2| \hat{e}_1 e^{i(\vec{k} \cdot \vec{r} - \omega t + \delta + \chi_2)} \right\} \\ &= \frac{c|k|}{\omega\mu} e^{-k_2 \hat{m} \cdot \vec{r}} \left[|E_1| \hat{e}_2 \cos(k_1 \hat{m} \cdot \vec{r} - \omega t + \delta + \chi_1) \right. \\ &\quad \left. - |E_2| \hat{e}_1 \cos(k_1 \hat{m} \cdot \vec{r} - \omega t + \delta + \chi_2) \right] \end{aligned}$$

Unless $\chi_1 = \chi_2$ we see that the components of $\vec{E}$ and $\vec{H}$ in directions $\hat{e}_1$ and $\hat{e}_2$ will oscillate out of phase with each other. Thus the directions of $\vec{E}$ and $\vec{H}$ will oscillate in time and space, as well as the amplitudes of $\vec{E}$ and $\vec{H}$. The direction of $\vec{E}$ and $\vec{H}$ is no longer fixed.