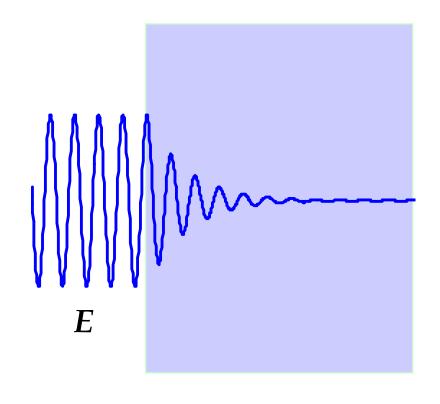
Today in Physics 218: conductors

- ☐ Electromagnetic waves in conductors
- ☐ Attenuation of the waves, and an electronic analogy
- ☐ Penetration of waves into conductors: skin depth



Electromagnetic waves in conductors

Consider a medium in which there is (still) no free electrical charge, but can be free currents:

$$J = \sigma E$$
 .

Then the Maxwell equations become

$$\nabla \cdot \mathbf{D} = 0 \qquad \nabla \cdot \mathbf{B} = 0$$

$$\nabla \times \mathbf{E} = -\frac{1}{c} \frac{\partial \mathbf{B}}{\partial t} \quad \nabla \times \mathbf{H} = \frac{4\pi}{c} \mathbf{J} + \frac{1}{c} \frac{\partial \mathbf{D}}{\partial t}$$

or, getting rid of the auxiliary fields with the assumption that the medium is (still) linear,

$$\nabla \cdot \mathbf{E} = 0 \qquad \nabla \cdot \mathbf{B} = 0$$

$$\nabla \times \mathbf{E} = -\frac{1}{c} \frac{\partial \mathbf{B}}{\partial t} \qquad \nabla \times \mathbf{B} = \frac{4\pi\sigma\mu}{c} \mathbf{E} + \frac{\varepsilon\mu}{c} \frac{\partial \mathbf{E}}{\partial t}$$

Let's make a wave equation for the electric field as we did before, by taking the curl of one of the curl equations:

$$\nabla \times (\nabla \times E) = -\frac{1}{c} \nabla \times \frac{\partial B}{\partial t}$$

$$\nabla (\nabla E) - \nabla^2 E = -\frac{1}{c} \frac{\partial}{\partial t} (\nabla \times B)$$

$$\nabla^2 E = \frac{\varepsilon \mu}{c^2} \frac{\partial^2 E}{\partial t^2} + \frac{4\pi \sigma \mu}{c^2} \frac{\partial E}{\partial t} \quad .$$

For one-dimensional propagation (plane waves, zero incidence, just like waves on a string), this is

$$\frac{\partial^2 \mathbf{E}}{\partial z^2} = \frac{\varepsilon \mu}{c^2} \frac{\partial^2 \mathbf{E}}{\partial t^2} + \frac{4\pi \sigma \mu}{c^2} \frac{\partial \mathbf{E}}{\partial t} .$$

☐ You many have seen this equation before in classical mechanics: it is the equation of motion of a damped harmonic oscillator. Veterans of F2002's PHY 217 have seen it used in connection with *LRC* circuits.

Let's solve it for some arbitrary component of E by separation of variables. Let $E_i = Z(z)T(t)$ and divide through by ZT:

$$\frac{1}{Z}\frac{d^2\mathbf{Z}}{dz^2} = \frac{1}{T}\left(\frac{\varepsilon\mu}{c^2}\frac{d^2T}{dt^2} + \frac{4\pi\sigma\mu}{c^2}\frac{dT}{dt}\right) = -k'^2 = \text{constant} .$$

For the *Z* part,

$$\frac{d^2\mathbf{Z}}{dz^2} = -k'^2 Z \quad \Rightarrow \quad Z = Ae^{\pm ik'z} \quad ,$$

as usual.

The *T* part is a little more interesting:

$$\frac{d^2T}{dt^2} + \frac{4\pi\sigma}{\varepsilon} \frac{dT}{dt} + \frac{c^2k'^2}{\mu\varepsilon} T = 0 \quad .$$

Try a solution of the form $T = Ce^{\alpha t}$, so that

$$\alpha^2 T + \frac{4\pi\sigma\alpha}{\varepsilon} T + \frac{c^2 k'^2}{\mu\varepsilon} T = 0 \quad .$$

Unless C = 0 (trivial solution),

$$\alpha^2 + \frac{4\pi\sigma\alpha}{\varepsilon} + \omega^2 = 0$$
 , $\omega^2 \equiv \frac{c^2 k'^2}{\mu\varepsilon}$

for which
$$\alpha = -\frac{2\pi\sigma}{\varepsilon} \pm \frac{1}{2} \sqrt{\left(\frac{4\pi\sigma}{\varepsilon}\right)^2 - 4\omega^2}$$

or, rearranged a little,

$$\alpha = -\frac{2\pi\sigma}{\varepsilon} \pm i\omega t \sqrt{1 - \frac{4\pi^2\sigma^2}{\varepsilon^2\omega^2}} \quad .$$

Thus the full solution to the wave equation has the form

$$E_i = \underbrace{Ae^{\pm ik'z\pm i\omega t\sqrt{1-4\pi^2\sigma^2/\varepsilon^2\omega^2}}}_{\mbox{Plane waves, propagating in either direction along z.}} \underbrace{e^{-2\pi\sigma t/\varepsilon}}_{\mbox{Attenuation (damping)}}$$

So travelling electromagnetic waves are damped out, eventually, in a conductor – attenuated exponentially, decreasing by a factor of e in a time constant of

$$\tau = \frac{\varepsilon}{2\pi\sigma} = \frac{\rho\varepsilon}{2\pi} \quad . \quad \rho = \text{resistivity}$$

Electronic analogy

What are we to make of this unlikely combination of resistivity ρ and dielectric constant ε ?

Consider a circular wafer made of very weakly conducting material with resistivity ρ and dielectric constant ε , with radius is r and thickness $\ell \ll r$, and with highly conductive, metallic electrodes covering the circular faces. Its

resistance and capacitance

are

$$R = \frac{\rho \ell}{A} = \frac{\rho \ell}{\pi r^2} ,$$

$$C = \frac{\varepsilon A}{4\pi \ell} = \frac{\varepsilon r^2}{4\ell} = \frac{\pi \varepsilon_0 \varepsilon_r r^2}{\ell} \text{ in MKS.}$$

Electronic analogy (continued)

- □ Suppose that the wafer is charged up with a battery with voltage V_0 , and that the battery is disconnected at t = 0.
- □ Since the resistance and capacitance + have the same voltage, they can be considered to be in parallel, so the charge on the

electrodes follows

$$\frac{q}{C} + IR = 0 ;$$

$$\frac{dq}{dt} + \frac{1}{RC}q = \frac{dq}{dt} + \frac{4\pi}{\varepsilon\rho}q = 0 \quad \text{Use } q = CV:$$

$$\frac{dV}{dt} + \frac{4\pi}{\varepsilon\rho}V = 0 .$$

Electronic analogy (continued)

or, since the electric field in the wafer is given by $V = -E\ell$,

$$\frac{dE}{dt} + \frac{4\pi}{\varepsilon \rho} E = 0 \quad .$$

In any case it can be integrated directly:

$$\frac{dE}{dt} + \frac{4\pi}{\rho \varepsilon} E = 0 \implies \int \frac{dE}{E} = -\frac{4\pi}{\rho \varepsilon} \int dt$$

$$\ln \frac{E}{E_0} = -\frac{4\pi t}{\rho \varepsilon}$$

$$E(t) = E_0 e^{-4\pi t/\rho\varepsilon} = E_0 e^{-t/\tau}$$
 , $\tau = \frac{\rho\varepsilon}{4\pi} = RC$.

Evidently, $\rho \varepsilon / 4\pi$ is the "RC time constant" of matter!

Evidently the wave equation for fields in conductors allows forms for plane waves *similar* to what we've been using. Suppose we want the (rightward-bound) plane wave solution to look like $\tilde{E} = \tilde{E}_0 e^{i\left(\tilde{k}z - \omega t\right)} \quad \text{so } k \text{ is like the usual wavenumber, etc.}$

Put into the wave equation, this gives us

$$\begin{split} \frac{\partial^2 \mathbf{E}}{\partial z^2} &= \frac{\varepsilon \mu}{c^2} \frac{\partial^2 \mathbf{E}}{\partial t^2} + \frac{4\pi\sigma\mu}{c^2} \frac{\partial \mathbf{E}}{\partial t} \quad , \\ -k^2 \tilde{\mathbf{E}}_0 e^{i(kz - \omega t)} &= -\omega^2 \frac{\varepsilon \mu}{c^2} \tilde{\mathbf{E}}_0 e^{i(kz - \omega t)} - i\omega \frac{4\pi\sigma\mu}{c^2} \tilde{\mathbf{E}}_0 e^{i(kz - \omega t)} \quad , \\ k^2 &= \omega^2 \frac{\varepsilon \mu}{c^2} + i\omega \frac{4\pi\sigma\mu}{c^2} = \mu\varepsilon \frac{\omega^2}{c^2} \left(1 + i\frac{4\pi\sigma}{\varepsilon\omega}\right) \quad . \end{split}$$

This expression for \tilde{k} differs from the corresponding expression for nonconducting media by the factor in parentheses:

$$1 + i\frac{4\pi\sigma}{\varepsilon\omega} = \sqrt{1 + \left(\frac{4\pi\sigma}{\varepsilon\omega}\right)^2}e^{i\theta} \quad \text{, where } \theta = \arctan\left(4\pi\sigma/\varepsilon\omega\right).$$

Thus

$$\tilde{k} = \sqrt{\mu\varepsilon} \frac{\omega}{c} \left(1 + \left(\frac{4\pi\sigma}{\varepsilon\omega} \right)^2 \right)^{1/4} e^{i\theta/2}$$

$$= \sqrt{\mu\varepsilon} \frac{\omega}{c} \left(1 + \left(\frac{4\pi\sigma}{\varepsilon\omega} \right)^2 \right)^{1/4} \left(\cos\frac{\theta}{2} + i\sin\frac{\theta}{2} \right) \equiv k + i\kappa \quad .$$

We can simplify this with some trig work. First:

$$\cos(\alpha + \beta) = \cos\alpha\cos\beta - \sin\alpha\sin\beta$$

$$\Rightarrow \cos\theta = \cos\frac{\theta}{2}\cos\frac{\theta}{2} - \sin\frac{\theta}{2}\sin\frac{\theta}{2} = \cos^2\frac{\theta}{2} - \sin^2\frac{\theta}{2}$$

$$\Rightarrow \cos\frac{\theta}{2} = \sqrt{\frac{1 + \cos\theta}{2}} \quad , \quad \sin\frac{\theta}{2} = \sqrt{\frac{1 - \cos\theta}{2}} \quad .$$

But

$$1 = \cos^2 \theta + \sin^2 \theta = \cos^2 \theta + \cos^2 \theta \tan^2 \theta$$

$$=\cos^2\theta(1+\tan^2\theta)$$
, so

$$= \cos^{2}\theta \left(1 + \tan^{2}\theta\right), \text{ so}$$

$$\cos\theta = \sqrt{\frac{1}{1 + \tan^{2}\theta}} = \sqrt{\frac{1}{1 + \left(\frac{4\pi\sigma}{\varepsilon\omega}\right)^{2}}}.$$

Thus,

$$k = \sqrt{\mu\varepsilon} \frac{\omega}{c} \left(1 + \left(\frac{4\pi\sigma}{\varepsilon\omega} \right)^2 \right)^{1/4} \sqrt{\frac{1 + \cos\theta}{2}}$$

$$=\sqrt{\mu\varepsilon}\frac{\omega}{c}\sqrt{\left(1+\left(\frac{4\pi\sigma}{\varepsilon\omega}\right)^{2}\right)^{1/2}}\sqrt{\frac{1+\left(1+\left(\frac{4\pi\sigma}{\varepsilon\omega}\right)^{2}\right)^{-1/2}}{2}}$$

$$= \sqrt{\frac{\mu\varepsilon}{2}} \frac{\omega}{c} \left(\left(1 + \left(\frac{4\pi\sigma}{\varepsilon\omega} \right)^2 \right)^{1/2} + 1 \right)^{1/2} .$$

Similarly,

$$\kappa = \sqrt{\frac{\mu\varepsilon}{2}} \frac{\omega}{c} \left(\left(1 + \left(\frac{4\pi\sigma}{\varepsilon\omega} \right)^2 \right)^{1/2} - 1 \right)^{1/2} .$$

The upshot of all this is that κ , the imaginary part of the complex wavenumber \tilde{k} , also represents attenuation of electromagnetic waves in conductors:

$$\tilde{\mathbf{E}} = \tilde{\mathbf{E}}_0 e^{i(\tilde{k}z - \omega t)} = \tilde{\mathbf{E}}_0 e^{-\kappa z} e^{i(kz - \omega t)} .$$

The electric field amplitude decreases by a factor of e with every distance $d = 1/\kappa$ the wave covers:

$$d = \frac{1}{\kappa} = \sqrt{\frac{2}{\mu \varepsilon}} \frac{c}{\omega} \left(\left(1 + \left(\frac{4\pi \sigma}{\varepsilon \omega} \right)^{2} \right)^{1/2} - 1 \right)^{-1/2}$$
Skin depth
$$= \sqrt{\frac{2}{\mu_{r} \mu_{0} \varepsilon_{r} \varepsilon_{0}}} \frac{1}{\omega} \left(\left(1 + \left(\frac{\sigma}{\varepsilon_{r} \varepsilon_{0} \omega} \right)^{2} \right)^{1/2} - 1 \right)^{-1/2}$$
in MKS .